

## RESEARCH ARTICLES

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# A study on the comparison of CSAMT and MT data

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Analysis of Controlled Source Audio Frequency Magneto Telluric (CSAMT) data is conventionally carried out using Cagniard resistivity under far field conditions. However, the near field response, which cannot be avoided in the field operations, differs considerably from the far field response. The discrepancy has been recognized the world over. In India the potential of CSAMT has not been exploited in its entirety. To bridge this gap on the Indian scene, a beginning has been made by studying the far field-near field discrepancies. The near field response in CSAMT for Vertical Magnetic Dipole (VMD) and Horizontal Electric Dipole (HED) is computed for homogeneous and two layer media. The critical frequency below which the near field conditions validity is studied, and the correction factor to make near field data look like far field data over a homogeneous medium discussed.

THE Controlled Source Audio-Frequency Magneto-telluric method (CSAMT)<sup>1,2</sup> is a frequency domain electromagnetic sounding technique with a fixed grounded electric dipole or horizontal loop (or a vertical magnetic dipole) as signal source. Stronger and stable signal, time invariance of the primary field polarization and its selectability confer a marked advantage for CSAMT over natural source methods (MT and AMT).

While two perpendicular horizontal components of electric and magnetic fields ( $E_x$  and  $H_y$ ) are measured in scalar CSAMT,  $E_x$ ,  $H_y$ ,  $H_x$  and  $H_z$  components are measured in vector CSAMT. Usually five components (namely  $E_x$ ,  $E_y$ ,  $H_x$ ,  $H_y$  and  $H_z$ ) with two source polarizations are measured in Tensor CSAMT.

CSAMT results have been reported over sulphide targets<sup>3</sup>, over massive sulphide body in Japan<sup>4</sup>, over lead/zinc/silver deposit in Alaska<sup>5</sup>, over uranium and sulphide/graphite targets<sup>4</sup>, over base metals in Finland<sup>6-8</sup>. CSAMT measurements have been carried out by Uchida *et al.*<sup>9</sup> over a copper/lead/zinc deposit in Japan. CSAMT measurements have also been proved successful in the search for gold deposits in Japan<sup>10</sup>.

CSAMT measurements have been carried out for geothermal exploration in USA<sup>12</sup>, in Japan<sup>4,13</sup> and in Hawaii<sup>14</sup>. In petroleum exploration, CSAMT has been used for structural mapping<sup>4,15-18</sup>.

CSAMT method proved effective in the investigations for groundwater quality<sup>19-23</sup>, for structural analysis in mine planning<sup>24</sup>, detecting voids in underground mines<sup>25</sup>, mapping burn fronts in underground coal gasification and coal mine fires<sup>24,26</sup>. However, practically no study has been reported from India.

### Far field and near field

Usually, if the transmitter receiver separation  $r \ll \delta$  and the induction number  $|kr| \ll 1$  (where  $|k| = (\sigma\mu\omega)^{1/2}$  is the wave number of quasistatic case) it is known as the 'near field' zone. On the other hand if  $r \gg \delta$ ,  $|kr| \gg 1$  it is known as 'far field' zone.

The theory for the MT, AMT, AFMAG and VLF case is well-formulated since the source independence of the EM field is assumed<sup>27</sup>. However, in CSAMT as the receiver-transmitter separation becomes less than 3 skin depths (skin depth,  $\delta = (2/\sigma\mu\omega)^{1/2}$  for quasistatic case) at low frequencies, the field becomes 'near field' rather

than 'far field', and the transmitter acts more like a local EM source than as a magneto-telluric source<sup>28</sup>. The finite source problem has been considered by a number of authors<sup>1,2,29-34</sup>.

A fairly detailed discussion of EM field component behaviour under these two asymptotic ranges is given by Kaufman and Keller<sup>34</sup>.

Here we assess the far field and near field CSAMT resistivity over homogeneous and layered media using Vertical Magnetic Dipole (VMD) and Horizontal Electric Dipole (HED).

### The study

The fundamental aspects of the electromagnetic fields in earth study consisting of Maxwell's equations, wave equation, solution for a homogeneous earth, quasistatic and dielectric limits, wave number, etc, being well documented, the initial development is not duplicated here. Similarly, the mathematical development of the solution of Helmholtz equation over layered media is also not considered here. Instead, we have confined ourselves to the main study of the CSAMT apparent resistivity in this article.

### Vertical magnetic dipole solution (homogeneous earth)

The EM field components of vertical magnetic dipole placed on the surface of the homogeneous earth can be written as<sup>31-37</sup>

$$E_\phi = (Mi\omega\mu/4\pi r^2)e_\phi, \quad (1)$$

$$H_r = (M/4\pi r^3)h_r, \quad (2)$$

$$H_z = (M/4\pi r^3)h_z. \quad (3)$$

$M$  is the magnetic moment of the dipole given by  $M = ISn$ ,  $I$  being the current in the loop,  $S$  the area enclosed by the loop, and  $n$  the number of turns.

The magnetic and electric numbers  $h_z$ ,  $h_r$  and  $e_\phi$  are introduced as<sup>34,37</sup>

$$h_z = \frac{2}{\xi^2} [9 - e^{i\xi} (9 - 9i\xi - 4\xi^2 + i\xi^3)]. \quad (4)$$

$$h_r = -i \frac{\pi}{2} \xi^2 \left[ J_2\left(\frac{\xi}{2}\right) H_2\left(\frac{\xi}{2}\right) - J_1\left(\frac{\xi}{2}\right) H_1\left(\frac{\xi}{2}\right) \right]. \quad (5)$$

$$e_\phi = \frac{2}{\xi^2} [-3 + e^{i\xi} (3 - 3i\xi - \xi^2)]. \quad (6)$$

where  $\xi = \bar{k}r$ ;  $\xi = |\xi|$ .

### Far field response ( $r \gg \delta$ ; $|kr| \gg 1$ )

In the far field, the magnetic and electric numbers can be written as<sup>27</sup>

$$h_z = -18i/|\xi^2|, \quad (7)$$

$$h_r = 6(1+i)/|\sqrt{2}\xi|, \quad (8)$$

$$e_\phi = 6i/\xi^2. \quad (9)$$

Thus in the far field, the real component of  $H_z$  and  $E_\phi$  are zero. The imaginary components of  $h_z$  and  $e_\phi$  decreases as  $1/r^2$ . The horizontal component  $H_r$  has equal real and imaginary components. The field components  $H_r$  and  $E_\phi$  decrease as  $1/r^4$  while the vertical field component  $H_z$  decreases as  $1/r^5$ . Thus in the far zone the field of a vertical magnetic dipole over a homogeneous earth remains practically horizontal and can be viewed as vertically incident plane wave. Solving for the far field wave impedance we have

$$Z = \frac{E_\phi}{H_r} = \sqrt{\frac{\mu\omega}{\sigma}} e^{-\frac{\pi}{4}},$$

and

$$\rho = \left( \frac{1}{\omega\mu} \right) \left| \frac{E_\phi}{H_r} \right|^2, \quad (10)$$

which is the same as the Cagniard's resistivity equation<sup>38</sup> employed in MT, AMT and VLF under plane wave assumption.

Further, the magnetic field of the vertical magnetic dipole placed over a homogeneous earth is elliptically polarized, with the plane of polarization lying in a vertical plane passing through the dipole moment. The magnetic numbers of the major and minor axes of the ellipse are given by<sup>36</sup>

$$h_a = 6/kr, \quad (11)$$

$$h_b = \left( \frac{18}{\sqrt{2}} \right) \left( \frac{1}{kr} \right)^2. \quad (12)$$

Thus on obtaining the ratio of the major and minor axes of ellipse of polarization it is possible to obtain the resistivity of the homogeneous earth as

$$\rho = (\mu\omega r^2/4) (h_a/h_b).$$

### Near field solution ( $r \ll \delta$ ; $|kR| \ll 1$ )

The near field asymptotic solutions for the magnetic and electric numbers can be written as<sup>36,37</sup>

$$h_z = -1 - \frac{2\sqrt{2}}{15} \xi^3 + i \frac{\xi^2}{4} \left( 1 - \frac{8\sqrt{2}}{15} \xi \right) \quad (13)$$

$$h_r = \frac{\xi^4}{16} \left( \ln \frac{\gamma \xi}{4} + \frac{1}{12} \right) + i \frac{\xi^2}{4} \left( 1 - \frac{\pi}{16} \xi^2 \right) \quad (14)$$

$$e_\phi = \left( 1 - \frac{\sqrt{2}}{15} \xi^3 \right) + i \frac{\xi^2}{4} \left( 1 - \frac{4\sqrt{2}}{15} \xi \right). \quad (15)$$

In view of the basic differences that exist between the 'far field' and 'near field' electromagnetic field component behaviour, significant discrepancies are introduced once the Cagniard resistivity (which is valid for MT and AMT and far field CSAMT) is computed for the near field measurements.

Figures 1 and 2 show the behavior of the CSAMT Cagniard apparent resistivity computed from equations (14) and (15) for near field measurements. The most common source effect is the distortion of apparent resistivity near the source. For a particular source-receiver separation  $r$  over a chosen ground resistivity, the computed resistivity approaches far field resistivity at a frequency designated by us as 'critical frequency'  $f_c$ . Below this frequency the apparent resistivity increases fairly linearly with decrease in frequency. Beyond this frequency, a distinct 'notch' or an 'under shoot', is recognized which is a decrease in resistivity below the true resistivity (Cagniard resistivity in case of homogeneous medium).

### Vertical magnetic dipole (two layer earth)

Over a two-layer medium the magnetic and electric numbers depend on  $q = 2h_1/r$ ,  $\beta = \sigma_1/\sigma_2$  and  $P_1 = |k_1 h_1|$ , where  $h_1$  and  $\sigma_1$  are the thickness and conductivity of the first layer and  $\sigma_2$  is the conductivity of the second layer.

The field due to a vertical magnetic dipole over a layered medium was studied extensively by several authors<sup>35,39-51</sup>. The near field solution ( $k_1 r \ll 1$ ,  $k_2 r \ll 1$ ) for small values of  $\beta$  are given by<sup>34,36,47</sup>

$$h_z = 1 + \left( \frac{P_1}{2} \right)^2 \left[ 1 + \left( \frac{\sigma_2}{\sigma_1} - 1 \right) \left( \frac{1}{(1+q^2)^{1/2}} \right) \right], \quad (16)$$

$$e_\phi = 1 - \left( \frac{P_1}{2} \right)^2 \left[ 1 + \left( \frac{\sigma_2}{\sigma_1} - 1 \right) (1+q^2)^{1/2} - q \right], \quad (17)$$

$$h_r = \left( \frac{P_1}{2} \right)^2 \left[ \frac{\sigma_2}{\sigma_1} - \left( \frac{\sigma_2}{\sigma_1} - 1 \right) \frac{q}{(1+q^2)^{1/2}} \right], \quad (18)$$

where  $P_1 = |k_1 r|$  and  $q = 2h_1/r$  as  $h_1 \rightarrow \infty$ ,  $q \rightarrow \infty$  and the above equations reduce to

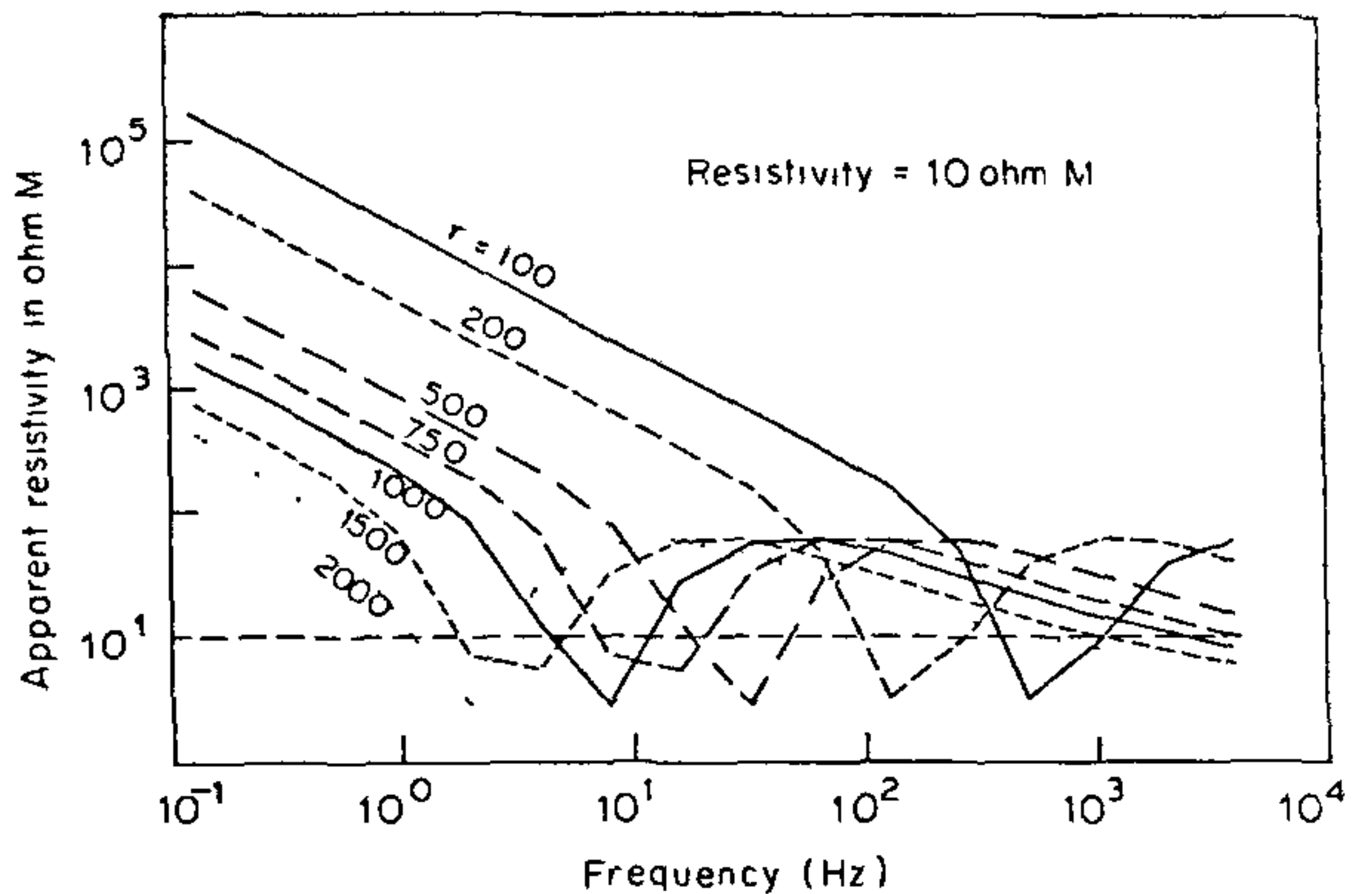


Figure 1. Homogeneous medium CSAMT response (VMD source)

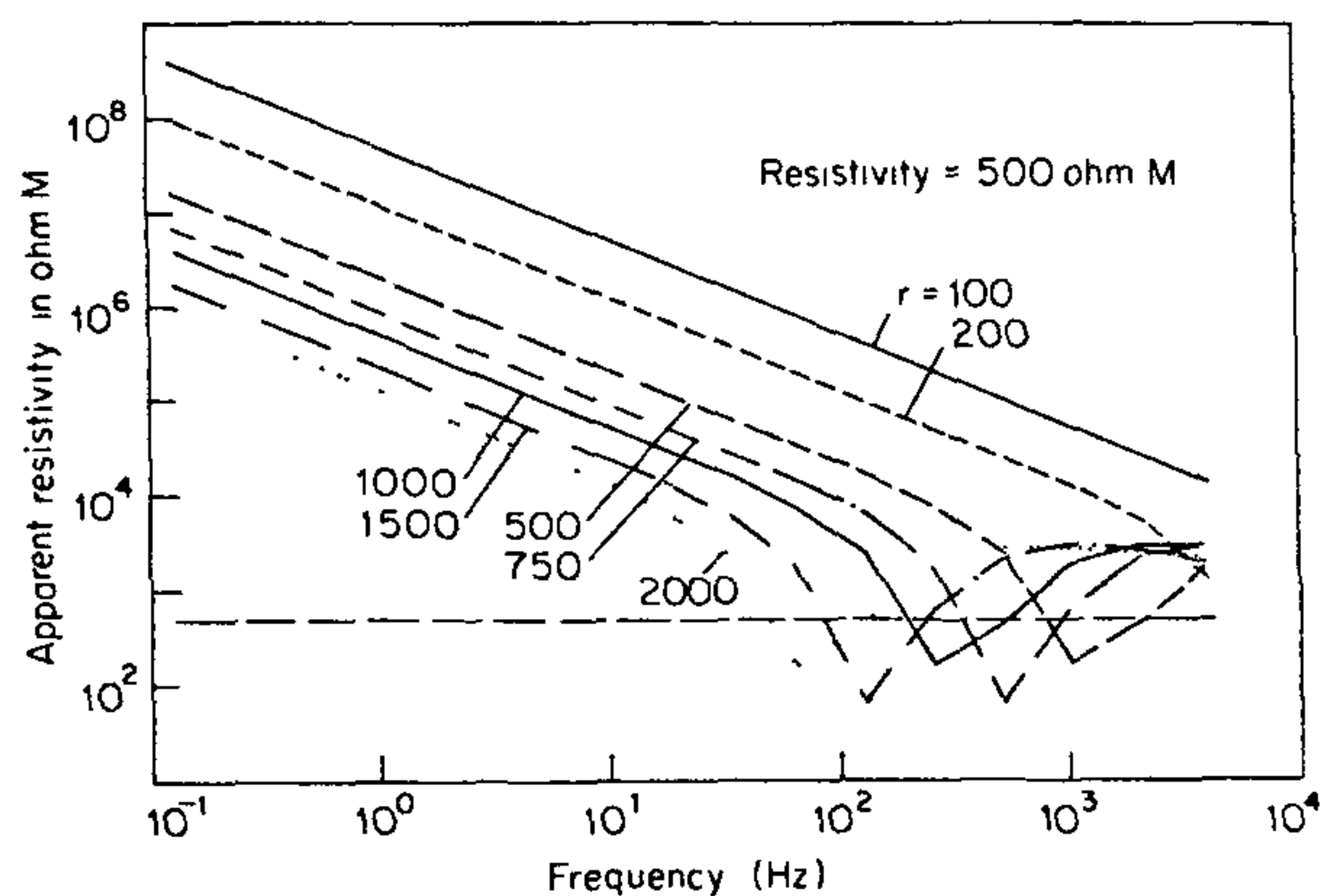


Figure 2. Homogeneous medium CSAMT response (VMD source)

$$e_\phi = 1 - (P_1/2)^2, \quad (19)$$

$$h_z = 1 + (P_1/2)^2, \quad (20)$$

$$h_r = (P_1/2)^2 (\sigma_2/\sigma_1). \quad (21)$$

The field components over a homogeneous medium [equations (1.153), (1.157) and (1.163) for  $n=4$  on pages 37, 38 and 40 of Kaufman and Keller<sup>34</sup>] coincide with equations (19) to (21).

In Figure 3, homogeneous medium Cagniard resistivity is compared with two-layer Cagniard resistivity for large values of  $h$ . The near field values fully agree with the homogeneous earth solution in its limiting case as  $h_1 \rightarrow \infty$ ,  $q \rightarrow \infty$ . However, beyond a certain frequency where the undershoots occur, the two solutions differ.

Figures 4 and 5 show the Cagniard resistivity over a two-layer medium for  $q$  and  $\beta$  values mentioned in the respective figures within the frequency band of 0.125 to 4096 Hz.

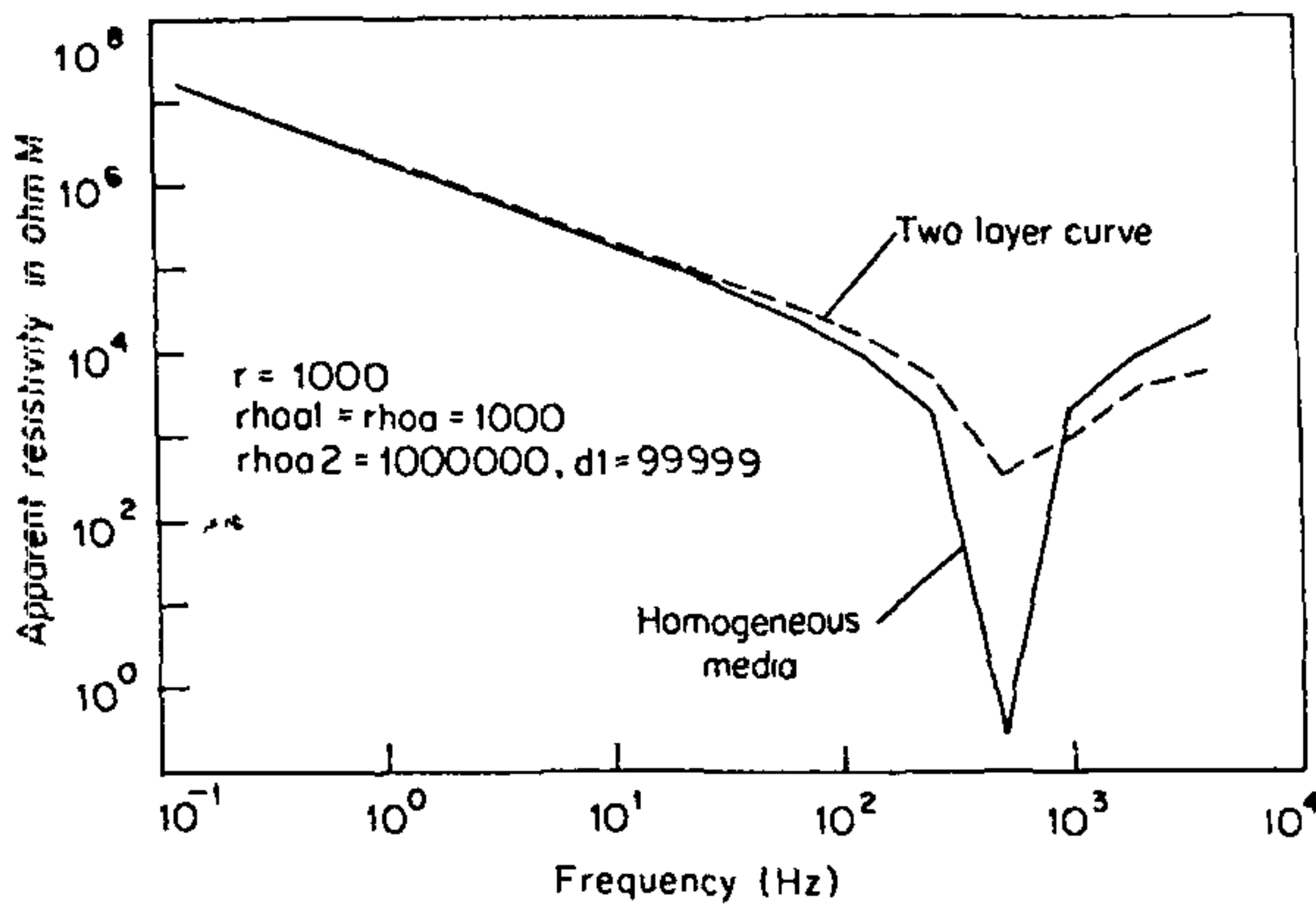


Figure 3. CSAMT response (VMD source)

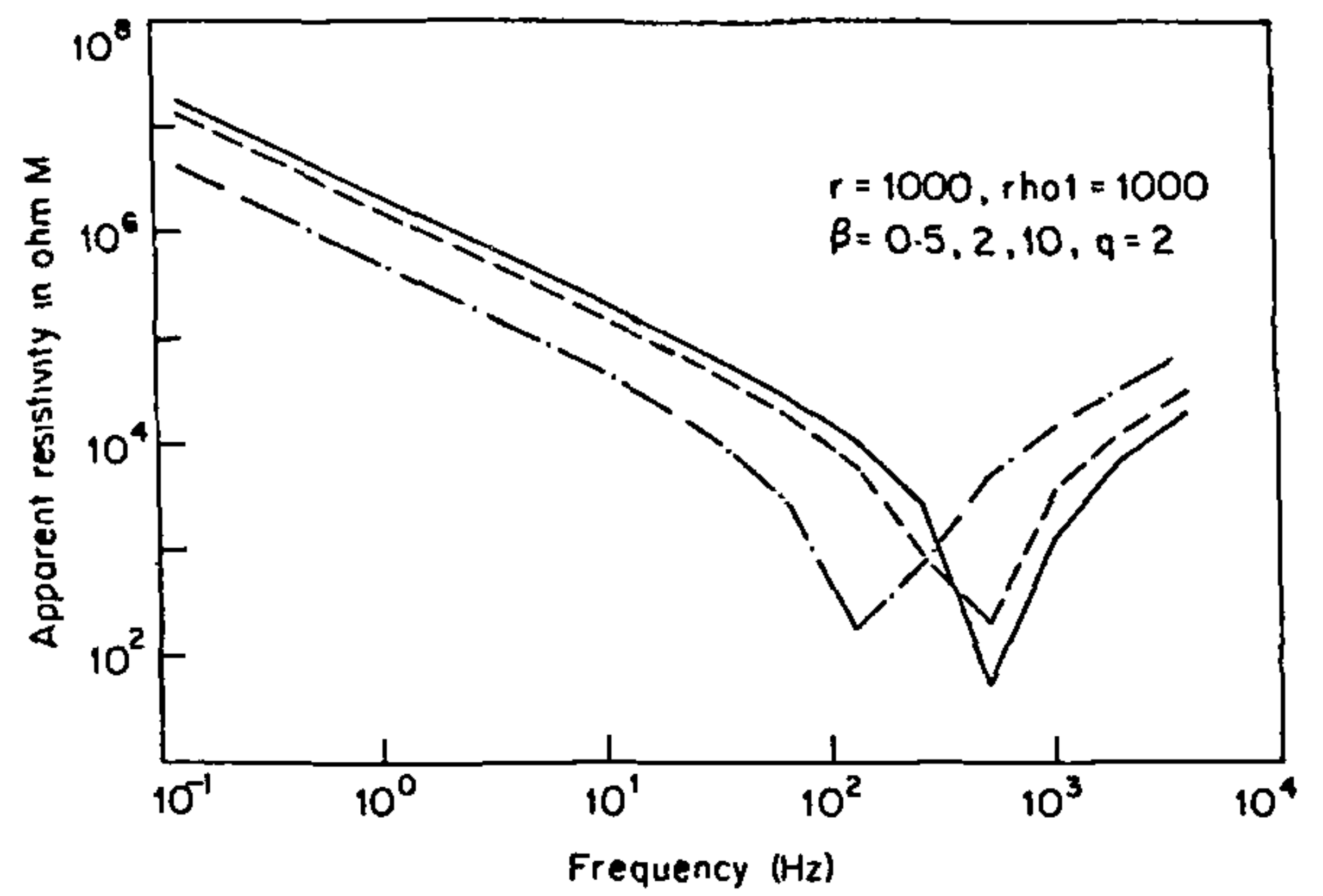


Figure 5. Two-layer medium CSAMT response (VMD source).

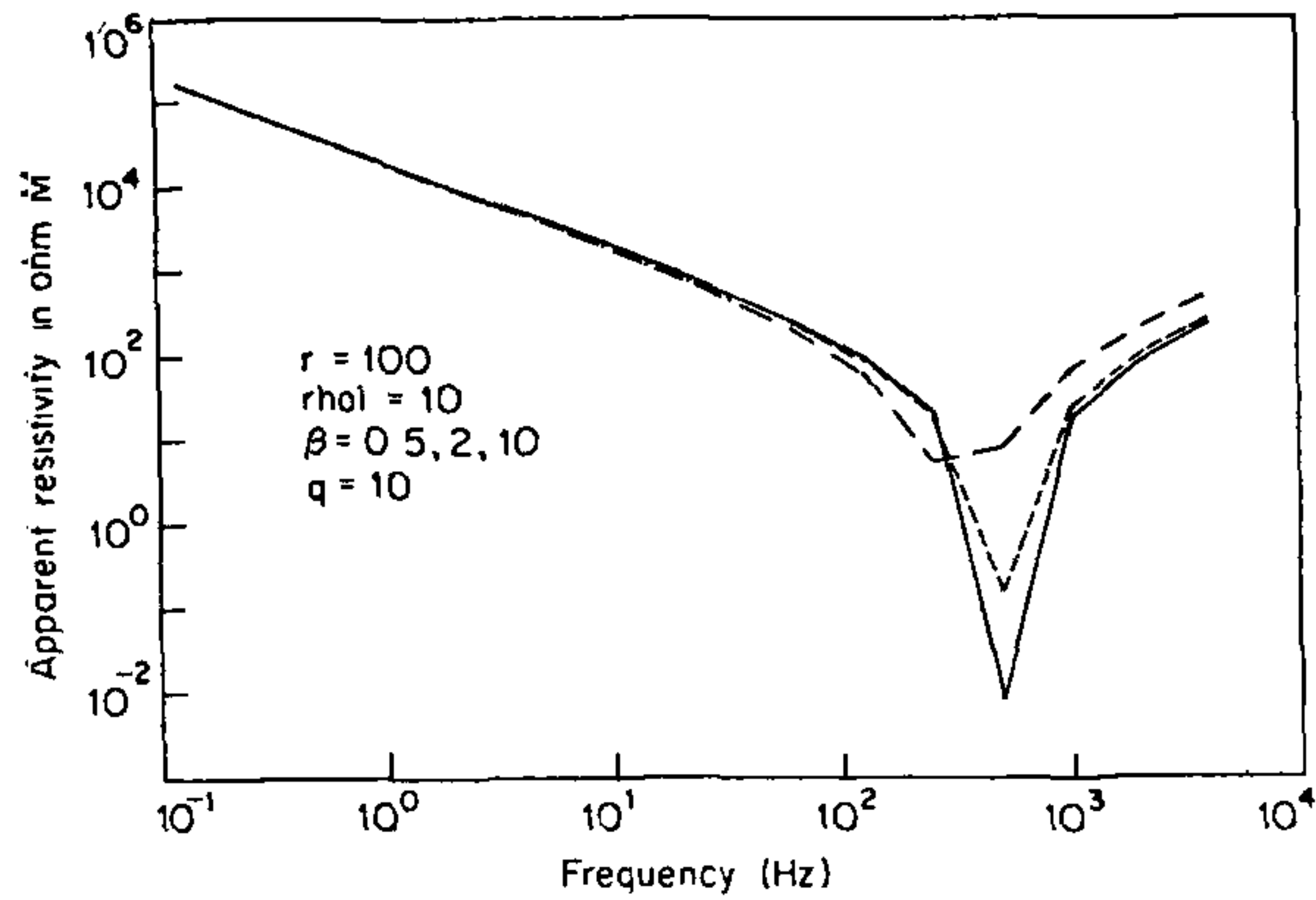


Figure 4. Two-layer medium CSAMT response (VMD source)

**Horizontal electric dipole solution (homogeneous earth)**

The electromagnetic fields due to grounded cable over a layered medium have been extensively studied by different authors<sup>30, 33-35, 48, 50-55</sup>.

The field components can be written as<sup>34, 37</sup>

$$h_x = \frac{xy}{r^2} i \frac{\pi}{2} \xi^2 \left[ J_1\left(\frac{\xi}{2}\right) H_2\left(\frac{\xi}{2}\right) + J_2\left(\frac{\xi}{2}\right) H_1\left(\frac{\xi}{2}\right) \right], \quad (22)$$

$$h_y = -i\pi \left\{ J_1\left(\frac{\xi}{2}\right) H_1\left(\frac{\xi}{2}\right) - \frac{\xi^2}{y} 2r^2 \times \left[ J_1\left(\frac{\xi}{2}\right) H_2\left(\frac{\xi}{2}\right) + J_2\left(\frac{\xi}{2}\right) H_1\left(\frac{\xi}{2}\right) \right] \right\}, \quad (23)$$

$$h_z = -2y/r \xi^2 [3 - e^{\xi} (3 - 3i\xi - \xi^2)], \quad (24)$$

$$e_x = 6(x^2/r^2) - 4 + 2e^{\xi} (1 - i\xi), \quad (25)$$

$$e_y = 6xy/r^2, \quad (26)$$

$$e_z = (-i\pi x/r) \xi^2 J_1(\xi/2) H_1(\xi/2), \quad (27)$$

where  $\xi = \bar{k}r$ ;  $\xi = |\xi|$ .

$J_n$  and  $H_m$  represent modified Bessel functions of the  $n$ th order. Introducing the electric and magnetic numbers for electric dipole as  $E_i = (P/4\pi r^3 \sigma_1) e_i$  and  $H_i = (P/4\pi r^2) h_i$ ,  $P = I \cdot dl$ , where  $dl$  is the length of the bipole and  $I$  is the current. The equations of the various components of the electrical and magnetic fields can be obtained from these equations by replacing the suffix  $i$  with the corresponding component (i.e.  $x$  or  $y$  or  $z$ ).

*Far zone response ( $r \gg \delta$  and  $|kr| \gg 1$ )*

The far zone components can be written as<sup>37</sup>

$$h_x = (6xy/r^2) (1 + i) / \sqrt{2} \xi, \quad (28)$$

$$h_y = \left( 4 - 6 \frac{x^2}{r^2} \right) \left( \frac{1+i}{\sqrt{2}\xi} \right), \quad (29)$$

$$h_z = 6i(y/r)(1/\xi^2), \quad (30)$$

$$e_x = 6(x^2/r^2) - 4, \quad (31)$$

$$e_z = - (3/2) (x/r), \quad (32)$$

where  $\xi = (\sigma\omega\mu r)^{1/2}$ .

The wave impedance in the far field zone is given by  $Z = E_x/H_y$ ,

$$Z = (i\omega\mu/2)^{1/2} = (\mu\rho\omega)^{1/2} \cdot e^{i\pi/4}$$

and

$$\rho = \left( \frac{1}{\omega\mu} \right) \left| \frac{E_\phi}{H_r} \right|^2, \quad (33)$$

which is the same as eq. (10) in the case of far field zone for vertical magnetic dipole. Thus the resistivities computed using the Cagniard equation using farfield solutions of either vertical magnetic dipole or horizontal electric dipole sources coincide with the plane wave solution used in MT, AMT and VLF.

Further the magnetic field is elliptically polarized and is practically horizontal in farfield zone.

*Near field solution ( $r \ll \delta, |kr| \ll 1$ )*

The near field solutions in Cartesian coordinate system are given by<sup>36,37</sup>

$$h_x = \frac{xy}{r^2} \left[ 2 + \frac{\xi^2}{64} \left( \ln \frac{\gamma\xi}{4} - \frac{1}{3} \right) + i \frac{\xi^2}{8} \left( 1 - \pi \frac{\xi^2}{32} \right) \right], \quad (34)$$

$$h_y = -1 + \pi \frac{\xi^2}{32} + i \frac{\xi^2}{8} \left( \ln \frac{\gamma\xi}{4} - \frac{1}{4} \right), \quad (35)$$

$$h_z = \frac{y}{r} \left[ 1 - \frac{\sqrt{2}}{15} \xi^3 + i \frac{\xi^4}{4} \left( 1 - \frac{4\sqrt{2}}{15} \xi \right) \right], \quad (36)$$

$$e_x = 6 \frac{x^2}{r^3} - 2 - \frac{\sqrt{2}}{15} \xi^3 + i \xi^2 \left( 1 - \pi \frac{\xi^2}{32} \right), \quad (37)$$

$$e_y = 6(xy/r^2), \quad (38)$$

$$e_z = \frac{x}{r} \left[ \frac{\xi^4}{8} \left( \ln \frac{\gamma\xi}{4} - \frac{1}{4} \right) - i \xi^2 \left( 1 - \pi \frac{\xi^2}{32} \right) \right]. \quad (39)$$

The Cagniard resistivity for collinear configuration computed using near field solution is presented in Figures 6 and 7 along with the farfield Cagniard resistivity.

The general behaviour of the non-plane wave effects on the resistivity computed agrees well with that of the vertical magnetic dipole fields.

From the above discussion it can clearly be observed that the non-plane wave effects for the 'near zone' in the CSAMT method cannot be neglected and there is a significant departure for the far field conditions of the MT solution.

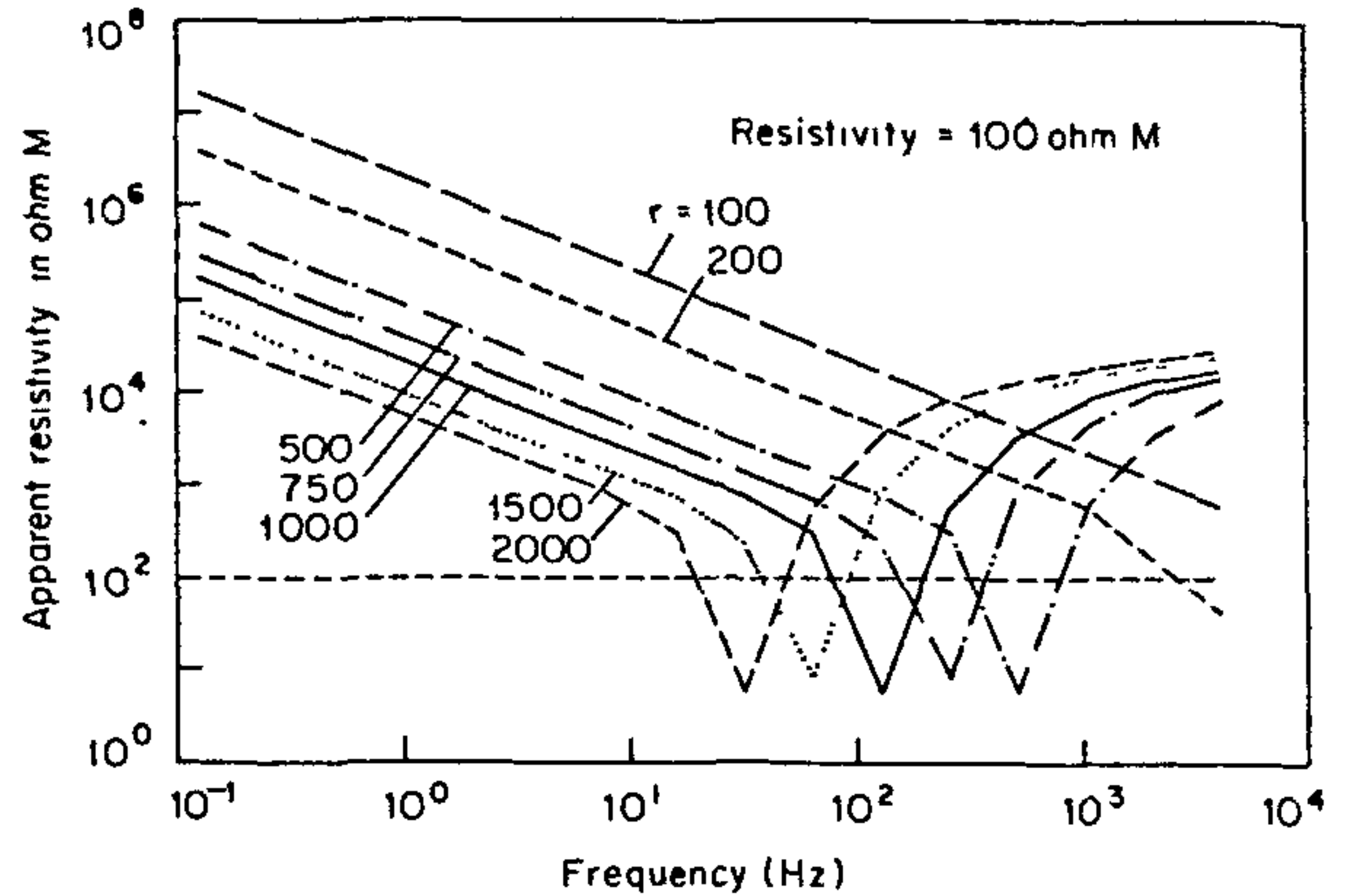


Figure 6. Homogeneous medium CSAMT response (HED source)

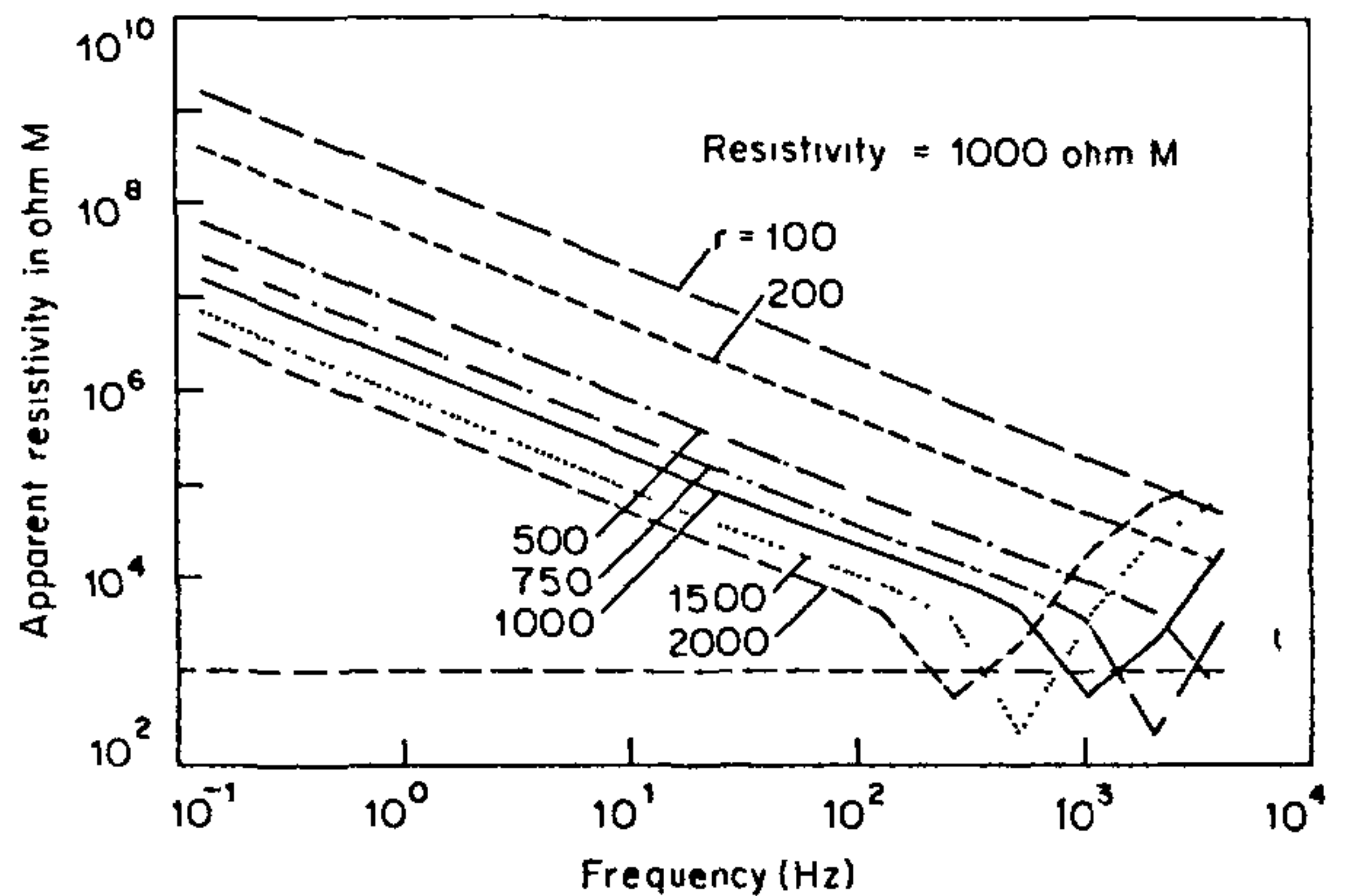


Figure 7. Homogeneous medium CSAMT response (HED source).

**Analysis**

*Critical frequency*

The present study is utilized to identify the 'critical frequency',  $f_c$  below which the near field solution for VMD and HED source is valid for different combinations of transmitter-receiver separation  $r$  and conductivity values  $\sigma$  of the homogeneous medium.

Figures 8 and 9 present nomograms indicating  $f_c$ .

*Near field corrections*

In the CSAMT survey, it is ideal to choose the  $f$  and  $r$  such that for all frequencies the measurements are made in the far field zone for a given geological situation. However, the distance between the transmitter and receiver locations is constrained in view of decreasing field strengths of different EM field components with

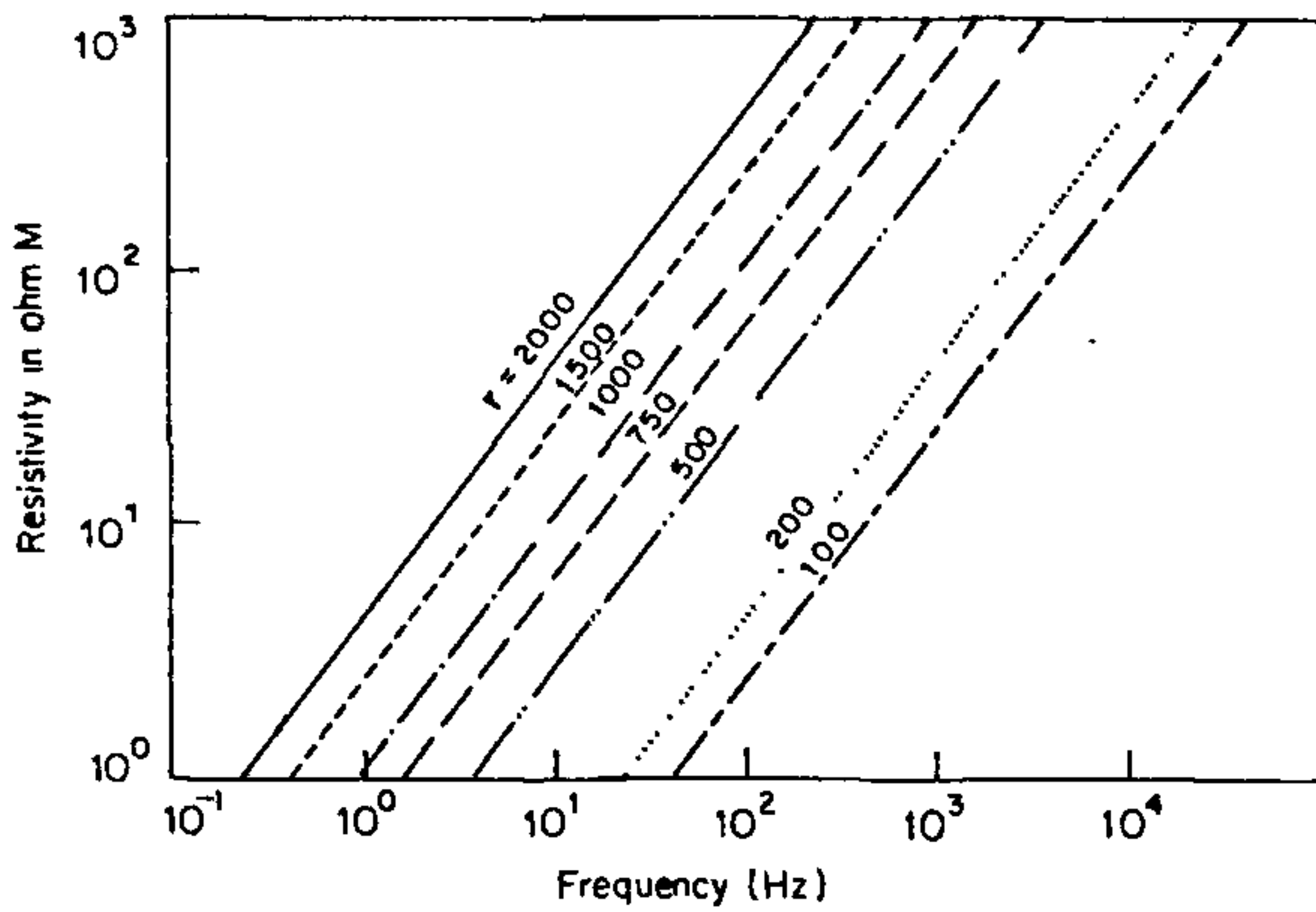


Figure 8. Critical frequency ( $f_c$ ) for magnetic dipole source

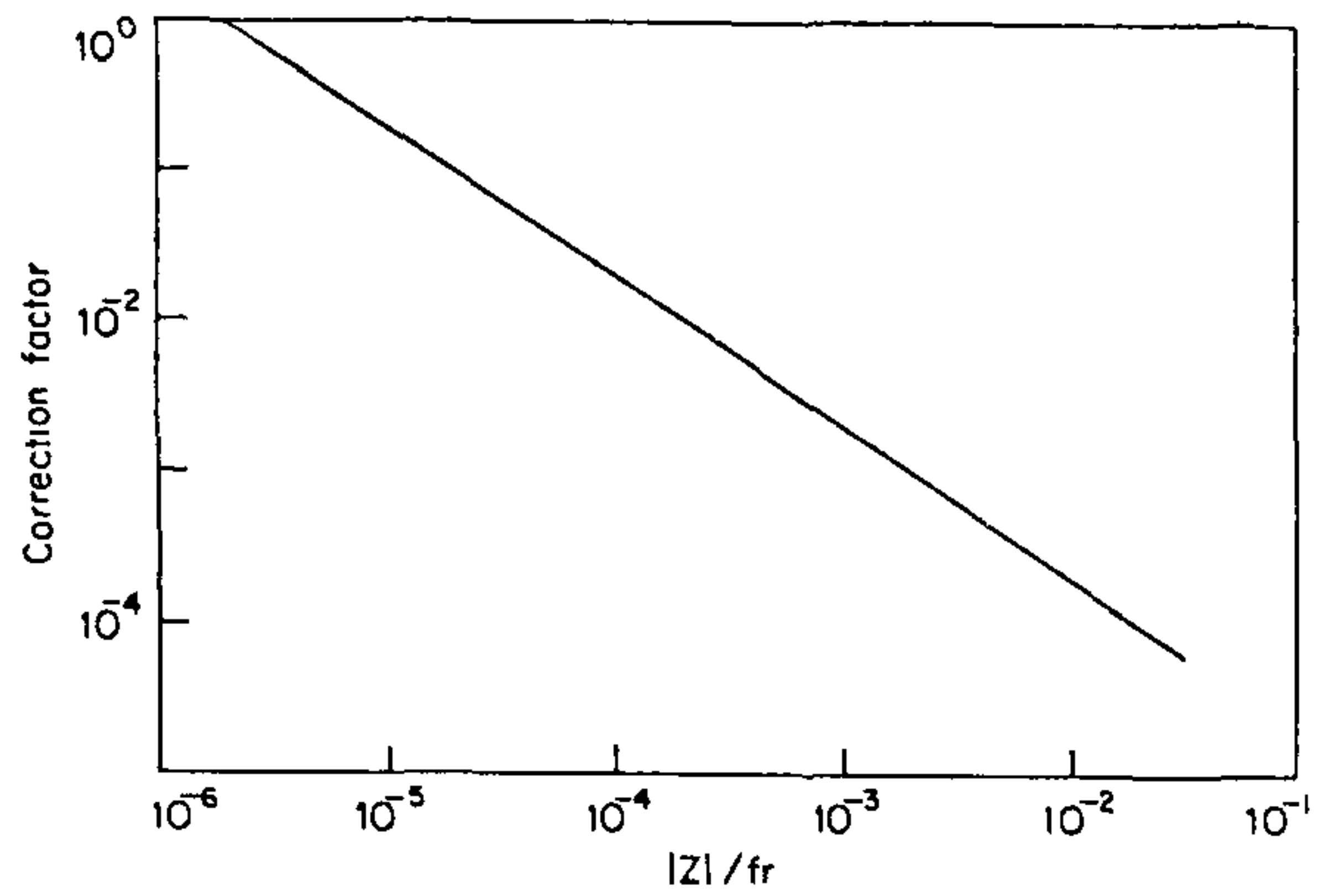


Figure 10. CSAMT near field correction ( $\eta$ ) for homogeneous medium

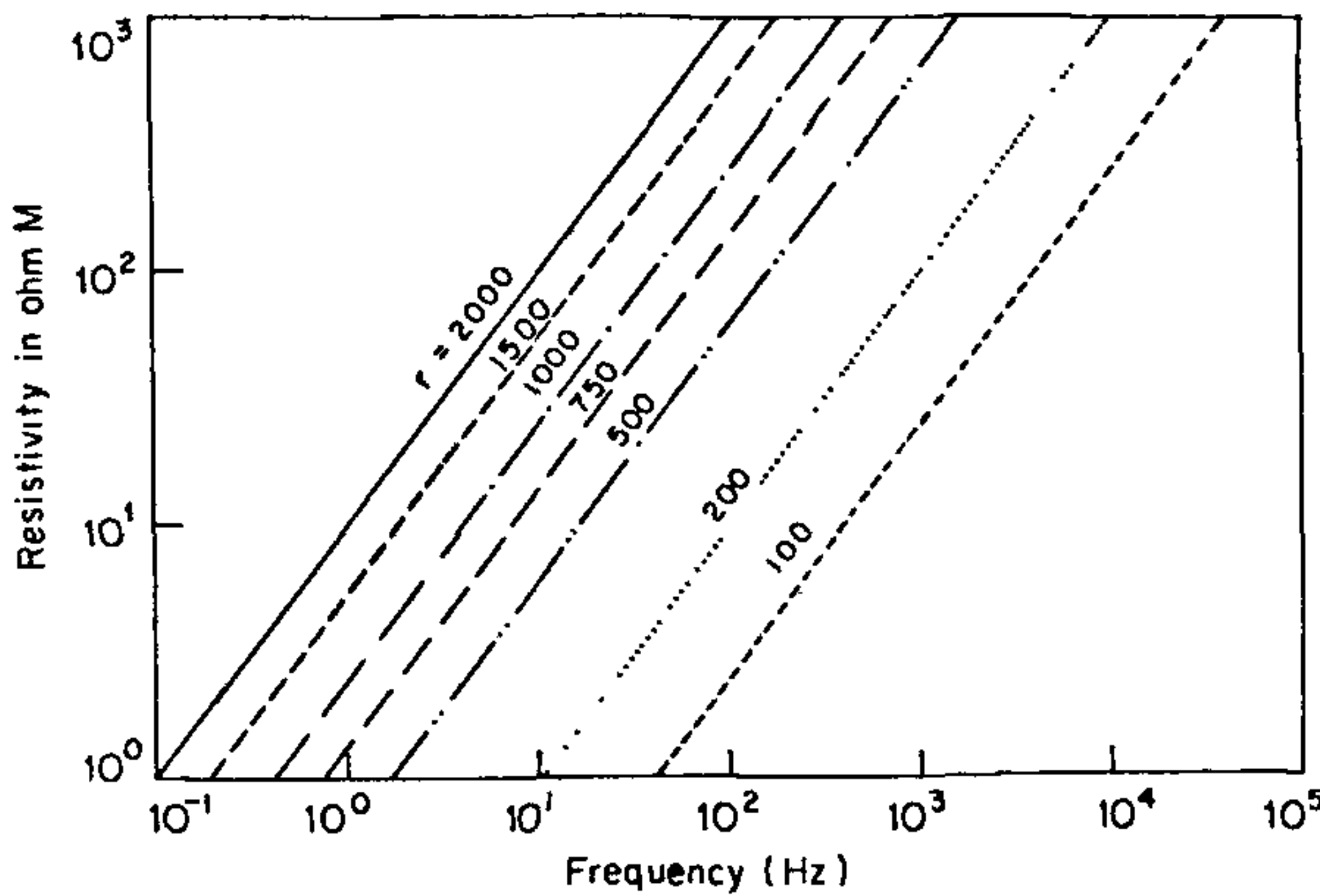


Figure 9. Critical frequency ( $f_c$ ) for electric dipole source.

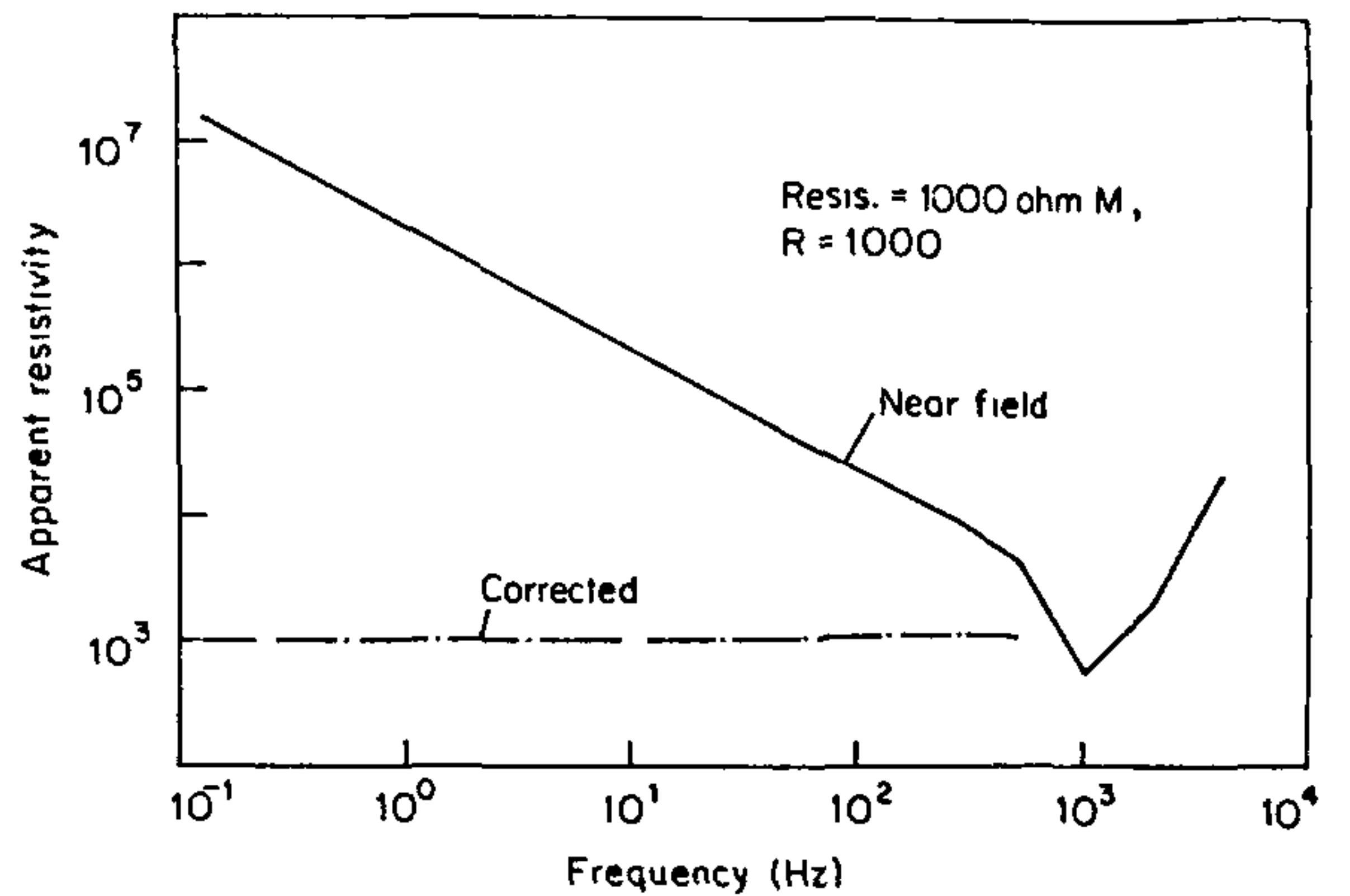


Figure 11. Near field correction (VMD source).

Table 1. Computation of corrected resistivity for CSAMT data

Frequency $f$ (Hz)	$\rho_{cs}$	$ Z /fr$	$\eta$	Corrected $\rho$
0.5	$4 \times 10^6$	0.008	0.00025	1000
1	$2 \times 10^6$	0.004	0.0005	1000
5	$4 \times 10^5$	0.0008	0.0025	1000
10	$2 \times 10^5$	0.0004	0.005	1000

distance (a detailed discussion is provided by Kaufman and Keller<sup>34</sup>, Svetov<sup>36</sup>, Vanyan<sup>50,51</sup>). Where the plane wave assumption is valid the field strength may be low and where the signal is strong (near field zone and transition zone  $r \approx \delta$ ,  $|kr| \approx 1$ ) the plane wave assumption may not be valid.

The near field correction curves for broad side position for  $r = 900$  m and  $dl = 600$  m and the frequency range of 1 to 10,000 Hz are produced by Bartel and Jacobson<sup>14</sup>. They have also presented an automated way of correcting the near field data utilizing the computed

apparent resistivities based on EMFIN4 computer programme<sup>56</sup> for homogeneous half space with resistivities  $\rho_1$  and  $\rho_2$ . The corrected  $\rho_c$  can be obtained as<sup>14</sup>

$$\log(\rho_c) = \frac{\log(\rho_m / \rho_{al}) \cdot (\rho_2 / \rho_1)}{\log(\rho_{an} / \rho_{al})} + \log \rho_1 \quad (40)$$

However, Maurer<sup>57</sup> and Szarkar<sup>58</sup> pointed out that such a correction is model specific and cannot be applied universally. MacInnes<sup>59</sup> describes an iterative technique which takes a far field Cagniard resistivity as a first guess to obtain the near field corrected apparent resistivity values.

The non-plane wave effects causing the departure of Cagniard resistivity in the near-field zone from the far field solutions can in principle be corrected on the basis of the field behaviour in these zones. The present study is utilized to obtain such a correction as described below.

Introducing a correction factor  $\eta$  such that  $\eta = \rho_{far}/\rho_{near}$  we find that  $\eta$  is a function of known quantities of  $Z (= E_{\phi}/H_r)$ ,  $f$  and  $r$ .  $\eta$  for  $\mu = \mu_0 = 4\pi \times 10^{-7}$  is presented in Figure 10.

Table 1 shows the scheme of such a correction for the  $\rho_{cs}$  shown in Figure 11. A similar correction factor was suggested by Yamashita and Hallof<sup>4</sup>.

The electric and magnetic fields in the near field solutions of electric and magnetic dipoles are independent of frequency and hence the impedance and the sounding data in the near field zone (which of course cannot be avoided in the field sounding data).

Thus the attention is to be diverted to transition zone data which contains more useful information but is complicated in view of the artificial 'undershoots' occurring in these zones.

In conclusion, it may be pointed out that any meaningful understanding of the CSAMT data can be achieved only by considering the full structure of the EM field due to a particular source, once the measurements involve near and transition zones.

## Conclusions

The results of the present study have been used to assess the parameters for which the near field conditions are valid.

However, there is a definite need to further investigate the CSAMT response using the general equations without resorting to approximations of the electromagnetic field and develop the CSAMT technique in the country in its totality.

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## MEETINGS/SYMPOSIA/SEMINARS

### International Symposium on Weed and Crop Resistance to Herbicides

Place Univ of Cordoba, Spain  
Date. 3-6 April 1995

Topics. The symposium will be structured in five sections each covering different aspects of herbicide resistance 1. Herbicide resistant weeds Historical cases, resistance in North America, Australia, Europe and other countries, Newly discovered resistant weeds, Monitoring and testing for resistance. 2 Herbicide resistance mechanisms. Altered target sites; Enhanced metabolism, Other mechanisms. 3. Genetics and biology of herbicide resistant weeds Genetics and heredity of resistance, Population dynamics, Other aspects of the biology of resistant weeds 4 Biotechnological approaches to the development of herbicide resistance in crops, problems and possibilities 5. Managing or avoiding herbicide resistance. Integrated mechanical, chemical and biological methods for weed control

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### International Conference on Differential Equations – Theory, Methods and Applications

Place Hyderabad, India  
Date 20-23 February 1995

Topics include: Ordinary, partial and functional (deterministic and stochastic) differential equations, Integral and integro-differential equations, Control theory, Difference equations, Dynamical systems (continuous and discrete), Ergodic theory and nonlinear waves, Mathematical biology, Neural networks and other aspects of mathematical modelling in physical, biological, engineering sciences and theoretical physics.

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### 2nd Agricultural Science Congress

Place Hyderabad  
Date. 19 - 21 January 1995

The following three symposia will be organized during the Congress 1 National Water Policy, 2 Vector Biology covering research and development aspects of the vectors of diseases of both farm animals and plants, 3 Integrated on-farm and off-farm employment

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